

# NOTES ON THE THEORY OF THE ALTERNATE CURRENT TRANSFORMER.

(A Paper read before the Sydney University Engineering Society, on June 20th, 1907).

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I feel that perhaps some explanation is needed as to why I should bring before this Society a subject which is treated of in ordinary text books on alternating currents and in special books on the same subject, but the fact is that I have found books very unsatisfactory when treating of this subject, and I think that by the methods that I have used, it can be made clearer than usual. Also I am convinced that many books are in error in at least one important point.

I have treated the subject all through by the use of vector or clock diagrams, and instead of simply drawing diagrams with no actual values, I have taken an imaginary transformer and worked out the actual values in each case.

In each case the results are checked by the following method:—The watts supplied to the transformer less the losses in the transformer should be equal to the watts given out, and, as will be seen from the diagrams, this is the case as nearly as can be expected. The primary and the secondary watts are always worked out from the voltage, current, and the power factor; while the losses are, except in one case, only the copper loss, which is easily determined. In one case, I introduced iron losses, to show their effect, but in all the other cases, the transformer has been taken as having no iron. This is done merely to simplify matters.

It will be noticed that no vectors to represent magnetic fluxes are included in the diagrams, as I am convinced that such lines are unnecessary and only complicate matters.

## 1.—LIST OF SYMBOLS USED IN THE PAPER.

- $f$  = Frequency in periods per second.
- $p$  =  $2 \pi f$ .
- $R_1$  = Resistance of primary of transformer in ohms.
- $R_2$  = Resistance of secondary of transformer in ohms.
- $n_1$  = Total number of primary turns.
- $n_2$  = Total number of secondary turns.
- $A$  = Cross sectional area of transformer core in square centimetres.
- $l$  = Mean length of magnetic circuit of transformer in centimetres.
- $E_1$  = E.M.F. supplied to the primary in volts.
- $E_2$  = E.M.F. available outside the secondary in volts.

- $E_{o2}$  = E.M.F. of secondary on open circuit in volts.  
 $E_{t2}$  = Total E.M.F. induced in the secondary in volts.  
 $C_{\mu}$  = Primary magnetising current in amperes on any load.  
 $C_{o\mu}$  = The same, no load.

Note.—This current is  $90^{\circ}$  in front of the counter E.M.F. due to self-induction.

- $C_{\eta}$  = Primary iron losses current in amperes.  
 $C_h$  = Primary hysteresis losses current in amperes.  
 $C_e$  = Primary eddy current losses current in amperes.

Note.— $C_{\eta} = C_h + C_e$  and  $C_o = \sqrt{(C_{\mu})^2 + (C_{\eta})^2}$  where

- $C_o$  = Primary no load current in amperes.  
 $C_1$  = Primary resultant current on load in amperes.  
 $C_2$  = Secondary current in amperes.  
 $C_2 \frac{n_2}{n_1}$  = Primary load current, *i.e.*, current which has to be supplied to the primary so that  $C_2$  may flow in the secondary.  
 $L_1$  = Co-efficient of self-induction of primary in Henrys.  
 $L_2$  = Co-efficient of self-induction of secondary in Henrys.  
 $M$  = Co-efficient of mutual induction of primary and secondary in Henrys.  
 $v$  = The leakage co-efficient between primary and secondary  
           =  $\frac{\text{total magnetic flux through primary due to } C_{\mu}}{\text{total magnetic flux through secondary due to } C_{\mu}}$   
 $k$  = The ratio of the total E.M.F. induced in the secondary to the counter E.M.F. in the primary due to  $C_{\mu}$ , which is, on open circuit, practically =  $\frac{E_1}{E_{t2}}$  for a transformer in which  $R_1$  is small.

- $+ \rho L_1 C_{\mu}$  = Counter E.M.F. of self-induction in primary due to  $C_{\mu}$ , in volts.  
                   { This E.M.F. is  $90^{\circ}$  behind  $C_{\mu}$ . }  
 $- \rho L_1 C_{\mu}$  = Impressed E.M.F. necessary in the primary to overcome  $+ \rho L_1 C_{\mu}$ .  
 $+ \rho L_1 C_2 \frac{n_2}{n_1}$  = Primary counter E.M.F. due to primary load current in volts.  
 $+ \rho L_2 C_2$  = Secondary counter E.M.F. due to secondary current in volts.  
 $+ \rho M C_2$  = Primary E.M.F. of mutual induction in volts, due to  $C_2$  in the secondary.  
 $+ \rho M C_2 \frac{n_2}{n_1}$  = Secondary E.M.F. of mutual induction in volts due to  $C_2 \frac{n_2}{n_1}$  in the primary.

$$\rho L_1 C_2 \frac{n_2}{n_1} - \rho M C_2 = \text{Primary counter E.M.F. in volts due to leakage.}$$

$$= \rho \left\{ L_1 - M \frac{n_1}{n_2} \right\} C_2 \frac{n_2}{n_1}$$

Note.—This E.M.F. is  $90^\circ$  behind the primary load current.

$$L^{1_1} = \left\{ L_1 - M \frac{n_1}{n_2} \right\}$$

= "Equivalent self-induction of primary" in Henrys.

$$- \rho L^{1_1} C_2 \frac{n_2}{n_1} = \text{E.M.F. which must be supplied from outside to overcome the counter E.M.F. due to leakage.}$$

Note.—This E.M.F. is  $90^\circ$  in front of the primary load current.

$$\rho L_2 C_2 - \rho M C_2 \frac{n_2}{n_1} = \text{Counter E.M.F. due to leakage in the secondary in volts.}$$

$$= \rho \left\{ L_2 - M \frac{n_2}{n_1} \right\} C_2$$

$$L^{1_2} = L_2 - M \frac{n_2}{n_1}$$

= Equivalent self-induction of secondary in Henrys.

$$- \rho L^{1_2} C_2 = \text{E.M.F. in volts used up in secondary because of leakage.}$$

$$L_1 = \frac{4\pi}{10} \frac{n_1}{l} A \frac{n_1}{10^8} \mu \text{ Henrys.}$$

$$L_2 = \frac{4\pi}{10} \frac{n_2}{l} A \frac{n_2}{10^8} \mu \text{ Henrys.}$$

$$M = \frac{4\pi}{10} \frac{n_1}{l} A \frac{n_2}{10^8} \times \frac{1}{v} \mu \text{ Henrys.}$$

$$M = \frac{1}{v} \sqrt{L_1 L_2} \text{ Henrys.}$$

$$E_{t2} = \rho M C \mu \text{ volts.}$$

$$= \rho /_1 C \mu \frac{n_2}{n_1} \times \frac{1}{v} \text{ volts.}$$

$\phi_0$  = Angle between  $E_1$  and  $C_0$ , open circuit.

$\phi_1$  = Angle between  $E_1$  and  $C_1$ , on load.

$\phi_{t2}$  = Angle between  $E_{t2}$  and  $C_2$ .

$\phi_2$  = Angle between  $E_2$  and  $C_2$ .

$W_1$  =  $E_1 C_1 \cos \phi_1$  in watts or  $E_1 C_0 \cos \phi_0$ .

$W_2$  =  $E_2 C_2 \cos \phi_2$  in watts.

$C^2 R$  = Total copper loss in transformer in watts.

## METHOD OF DRAWING THE DIAGRAMS.

In each case except VI, IX and X, the following are taken as known :—

The voltage supplied to the primary = 100 volts =  $E_1$ .

The secondary current =  $C_2$ .

The angle between the secondary current and the total secondary voltage  $E_{t2}$ .

The following is the method :—

1. When there is no leakage.

(a) Draw a vertical line to show the direction of  $-p L_1 C_\mu$  and  $E_{t2}$ .

(b) Take a point on it about the centre of the sheet and mark off  $C_\mu$  horizontally towards the left.

(c) Draw a line of the required length to represent  $C_2$  at the given angle from  $E_{t2}$ , starting from O.

(d) Produce this line through O and mark off on it  $C_2 \frac{n_2}{n_1}$  to the same scale of amperes as  $C_\mu$ .

(e) Find the resultant of  $C_\mu$  and  $C_2 \frac{n_2}{n_1} = C_1$

(f) On this resultant, starting from O, mark off  $C_1 R_1$  to the same scale of volts as  $E_1$ .

(g) Through the upper end of  $C_1 R_1$  draw a perpendicular line.

(h) With centre O and radius 100 volts describe an arc of a circle to cut this perpendicular in B.

(i) Draw O B<sub>1</sub> which is the position of  $E_1$  on this load.

(j) From B draw a line parallel to  $C_1 R_1$  to cut the vertical line through O at D. Then OD is  $-p L_1 C_\mu$ .

(k) Perpendicularly downwards from O is  $E_{t2} = -p L_1 C_\mu \times \frac{n_2}{n_1}$

(l) On  $C_2$  mark off  $C_2 R_2$ .

(m) Join the lower end of  $C_2 R_2$  to the lower end of  $E_{t2}$ . This line is  $E_2$ .

2. When there is leakage.

As in 1 for a, b, c, d, but leave the length of  $C_\mu$  unmarked.

(e) Through O draw a line at right angles to  $C_2$  and mark off on it to the primary and secondary scales of volts respectively  $-p L_1^1 C_2 \frac{n_2}{n_1}$  and  $-p L_2^1 C_2$ , each being 90° in front of its current.

(f) From the outer end of  $-p L_1^1 C_2 \frac{n_2}{n_1}$  draw a line parallel to  $C_2 \frac{n_2}{n_1}$  and equal to  $C_2 \frac{n_2}{n_1} R_1$ .

- (g) Through the end of  $C_2 \frac{n_2}{n_1} R_1$  draw a perpendicular line.  
 (h) With centre O and radius  $E_1$  draw an arc of a circle to cut this perpendicular in F.  
 (i) Join O.F. which is the position of  $E_1$  on this load.

Note.—This is not quite correct as it omits  $C\mu R_1$  which has yet to be determined but is very small.

- (j) The length of this perpendicular gives  $-p L_1 C\mu$ .  
 (k) Determine  $C\mu$  from the length of  $-p L_1 C\mu$  and mark it off.  
 (l) Find the resultant of  $C\mu$  and  $C_2 \frac{n_2}{n_1} = C_1$ .  
 (m)  $E_{t2}$  is  $-p L_1 C\mu \times \frac{n_2}{n_1} \times \frac{1}{v}$ .  
 (n) Mark off  $C_2 R_2$  on the  $C_2$  line from O.  
 (o) Through the lower end of  $E_{t2}$  draw a line parallel to  $-p L_1 C_2$  and equal to it.  
 (p) From the end of  $C_2 R_2$  draw a line to the end of the one drawn in (o).

This last drawn line is  $E_2$  on this load.

Details of imaginary transformer dealt with in the paper.

A = 100 square centimetres.

$n_1$  = 10,000 turns.

$n_2$  = 1,000 turns.

$l$  = 100 centimetres.

$R_1$  = 2.0 ohms.

$R_2$  = .020 ohms.

$f$  = 50 v.

$E_1$  = 100 volts,

Note.—Except in one case, No. II, the transformer is considered to have an endless magnetic circuit, and to contain no iron, consideration of that being omitted for the sake of brevity.

$$L_1 = \frac{4\pi}{10} \times \frac{100 \times (10^4)^2}{100 \times 10^8} = 1.25 \text{ Henry.}$$

$$p = 6.28 \times 50 = 314.$$

As there is no iron  $C\eta = 0$  and so  $C_0 = C\mu$ .

$$C\mu = \frac{100}{\sqrt{4 + (.314 \times 1.25)^2}} = \frac{100}{\sqrt{4 + (.392)^2}} = 255 \text{ amperes.}$$

$$p L_1 C\mu = 314 \times 1.25 \times .255 = 100 \text{ volts.}$$

$$\phi_0 = \tan^{-1} \frac{p L_1}{R_1}$$

$$= \tan^{-1} \frac{314 \times 1.25}{2.0} = \tan^{-1} 196 = 89.7^\circ$$

The watts lost in the primary at no load

$$\begin{aligned} &= 100 \times .255 \times \cos 89.7^\circ \text{ or } (.255)^2 \times 2.0 \\ &= 25.5 \times .0052 &= .065 \times 2.0 \\ &= .13 &= .13 \end{aligned}$$

If there is no magnetic leakage or  $v = 1$

$$\begin{aligned} M &= 1.25 \times 10^4 \times 10^3 \times 10^2 \times 10^{-2} \times 10^{-8} \\ &= .125 \text{ Henry.} \end{aligned}$$

$$E_{02} = 314 \times .255 \times .125 = 10 \text{ volts.}$$

$$E_1 \times \frac{n_2}{n_1} = 100 \times \frac{10^3}{10^4} = 10 \text{ volts.}$$

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#### DESCRIPTION OF THE DIAGRAMS.

I. This is simply the transformer, without magnetic leakage, on open circuit.

II. This diagram serves two purposes.

FIRSTLY.—It shows the effect of the iron losses, which are given quite arbitrary values.

In working out the iron losses, the total iron losses current, taken as usual, as being in phase with  $-p L_1 C\mu$  is multiplied by the voltage and by the cosine of the angle between them, which is  $1^\circ$ .

SECONDLY.—It shows why a transformer which has no leakage acts as a non-inductive circuit to the secondary current and the primary load current.

The secondary current  $C_2$  produces in the secondary a counter E.M.F. of self-induction  $= p L_2 C_2$ , but this is neutralised by an E.M.F. of mutual induction  $= p M C_2 \frac{n_2}{n_1}$  produced by the primary working current  $= C_2 \frac{n_2}{n_1}$

Similarly in the primary, there is a counter E.M.F.  $= p L_1 C_2 \frac{n_2}{n_1}$  but it is neutralised by an E.M.F. of mutual induction  $= p M C_2 \frac{n_2}{n_1}$

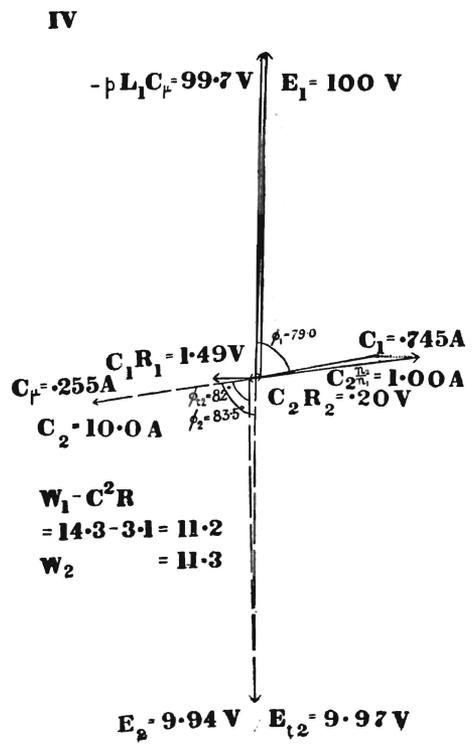
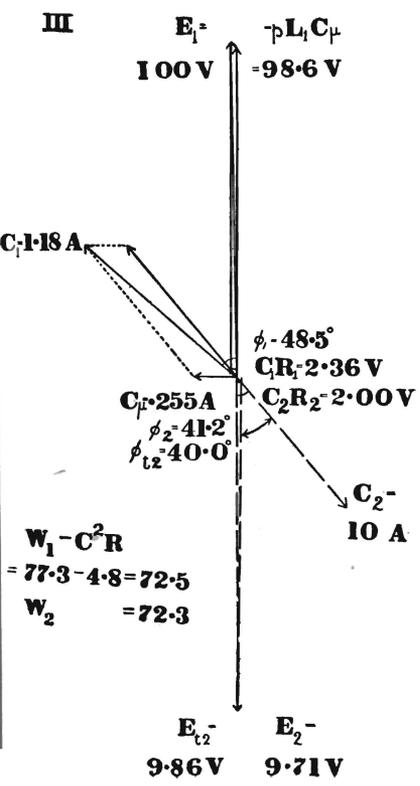
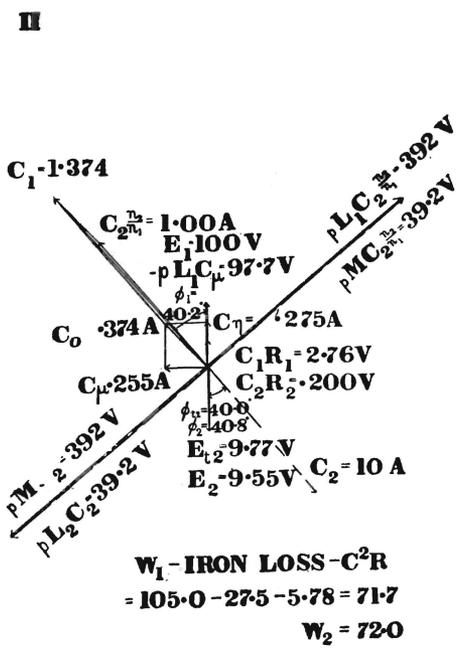
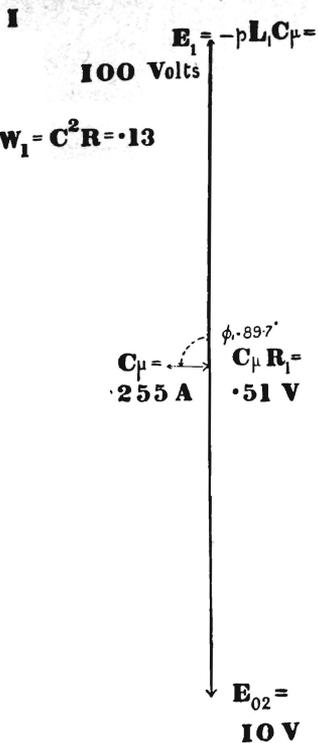
III. Shows a transformer without leakage working on an inductive load.

IV. Shows a transformer working with a leading secondary current.

V. Shows a transformer working on short circuit, where there is no leakage. In this case the voltage is reduced in the primary to that necessary to get the same secondary current of 10 amperes, as in the other diagrams. It will be noticed that  $C\mu$  is negligibly small.

VI. Shows a transformer, with  $v = 1.1$  working on an inductive load.

It will be noticed that  $E_{02}$ , the secondary voltage on open circuit, is no longer  $= p L_1 C\mu \times \frac{n_2}{n_1}$  but is  $p L_1 C\mu \frac{n_2}{n_1} \times \frac{1}{v}$  or  $p L_1 C\mu k$ .



There are two points about this diagram which demand attention.

- (a) The first is that as  $p L_1 C_\mu$  is now considerably smaller on load than an open circuit,  $C_\mu$  should be proportionately reduced. This has been done in the diagram, but it is a point neglected in most books.

Really  $C_\mu$  should be reduced in all cases, because  $-p L_1 C_\mu$  is reduced by the resultant along its line, of  $C_1 R_1$ , but this is generally very small, and, in the cases where there is no leakage this has been neglected.

- (b) The second is that in the secondary  $p L_2 C_2$  is now larger than  $p M C_2 \frac{n_2}{n_1}$  so that the secondary has some self-induction, and there is in it an E.M.F. of self-induction  $= p L_2 C_2$  where  $L_2$  is the "equivalent self-induction"  $= L_2 - M \frac{n_2}{n_1}$  and there is similarly on "equivalent self-induction" in the primary  $L_1 = L_1 - M \frac{n_1}{n_2}$

- (c) The primary working current is not  $C_2 \frac{1}{k}$  but  $C_2 \times \frac{n_2}{n_1}$ . It is generally given as  $C_2 k$ , but cannot be as  $p L_1 C_2 \frac{1}{k} = 357$  which is the same as  $p M C_2$ , when there would be no counter E.M.F. due to leakage in the primary.

And similarly in the secondary  $p L_2 C_2$  is 39.2 just as  $p M C_2 \frac{1}{k}$  is.

That my method of drawing this diagram is right, is shown by the usual test  $W_1 - C_2 R = W_2$ , when there is only a small error of .25 in 44.05 or .56 %.

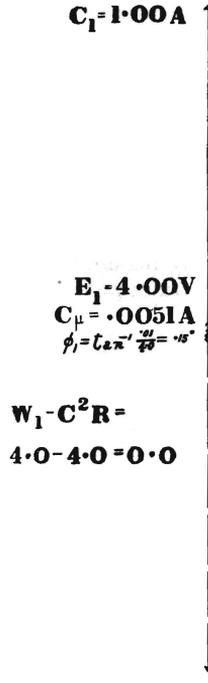
- VII. Shows the ordinary method, as given in books, of drawing this same diagram and, in this case alone amongst the set,  $W_1 - C_2 R$  is not equal to  $W_2$ . The error is 3.3 in 40 = 8.25 %.

There was a difficulty in drawing this as the books do not give any method of finding the primary and secondary counter E.M.Fs. due to leakage.

Prof. Sylvanus Thompson, for instance, merely says (Dynamo Electric Machinery, 1896 edition, p. 710), "we must take into account the unbalanced self-induction (if any) etc.

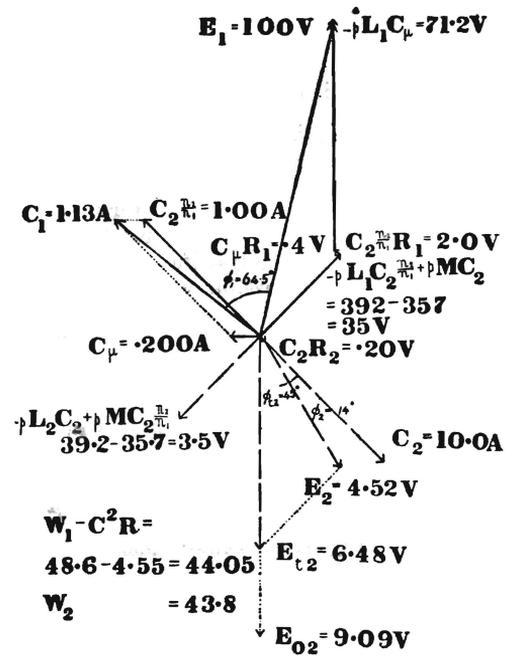
C. E. Lamb (Alternating Currents) gives a method, but it is very cumbersome, and his is the only book in which I have found a correct diagram (p. 64).

V

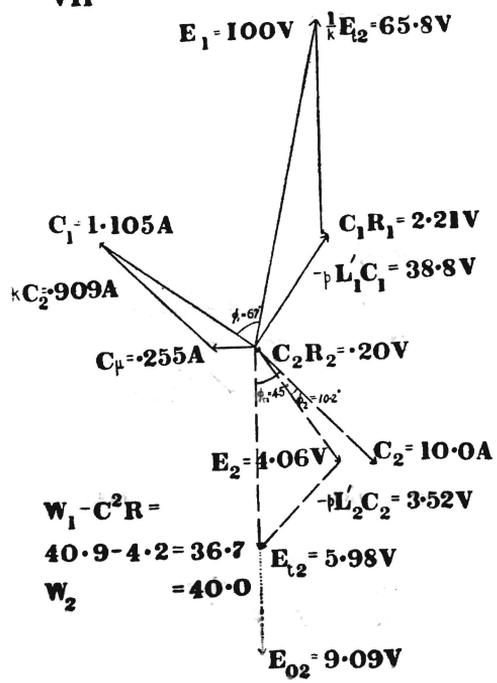


$C_1R_1 = 2.00V$   
 $\rightarrow L_1C_\mu = 2.00V$   
 $E_{t2} = C_2R_2 = .200V$

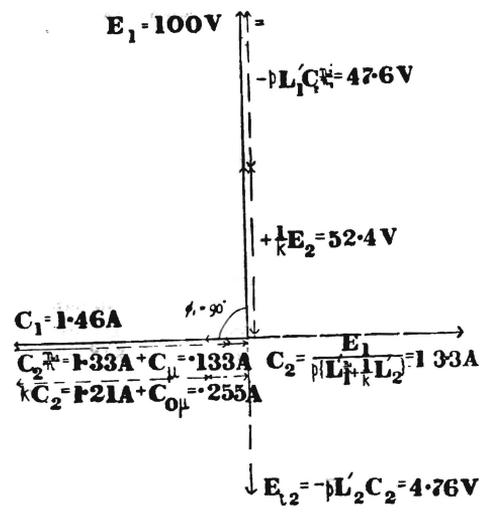
VI



VII



VIII



For the reason given in VI. (c) above, if the primary working current is taken as  $C_2 \frac{1}{k}$ , there is no counter E.M.F. due to leakage, and yet this is the value usually taken.

So I simply took  $L_1$  and  $L_2$  as the same values as in VI.

VIII. Shows the transformer working on short circuit with leakage but with no resistance. The secondary angle of lag inside the transformer is then  $90^\circ$  and the current is the maximum possible with  $E_1 = 100$  volts.

In this case it should be noted that:—

(a)  $C_1 = C_2 \frac{n_2}{n_1} + C_\mu$ , where  $C_\mu$  is the reduced value.

(b)  $C_1 = k C_2 + C_0\mu$ , where  $C_0\mu$  is the no load value.

(c)  $C_1 = k C_\mu$  where  $C_\mu$  is the reduced value.

IX. Is the same transformer with leakage and resistance, working on short circuit with the normal secondary current of 10 amperes. The secondary angle of lag, inside the transformer, is then fixed as its tangent must be  $\frac{p L_2^2}{R_2}$

X. Shows a simplified diagram, everything being transferred to the primary side. A method which can be adopted in each of the other cases.  $C_\mu$  is neglected in this diagram, being very small, and the primary part of the drawing is the triangle which Dr. D. R. Morris and Mr. G. A. Lister (Journal Inst. E.E., vol. 37, p. 265), call the "Characteristic Triangle" of the transformer.

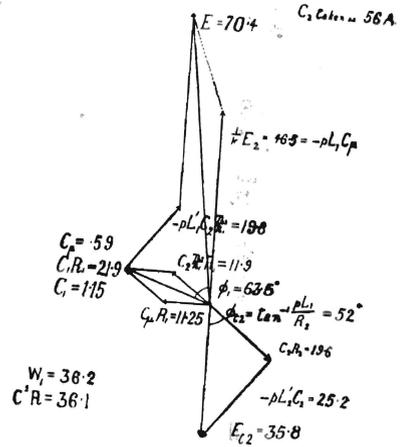
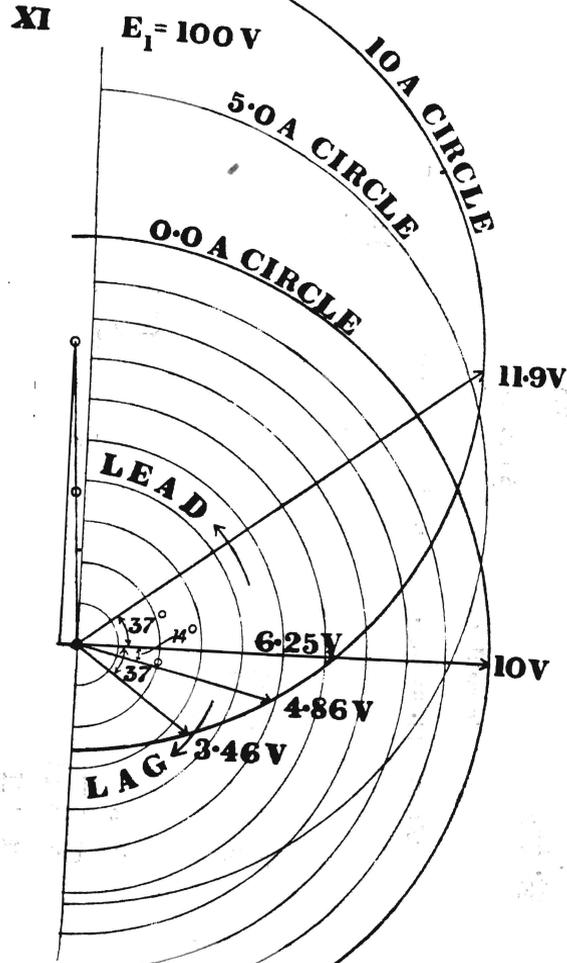
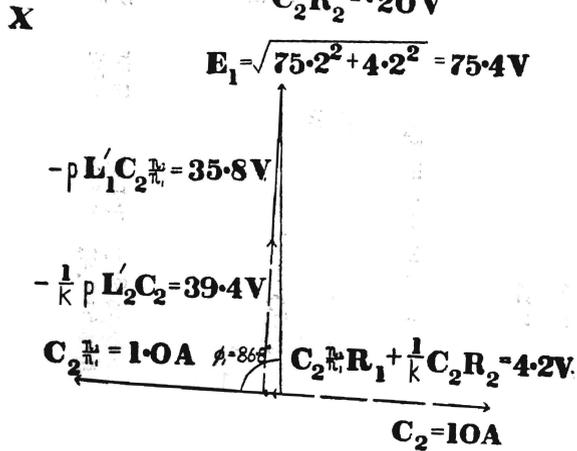
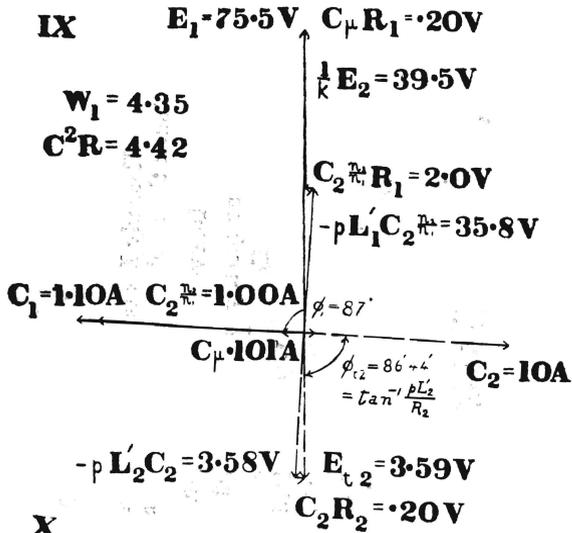
XI. Shows how the characteristic triangle can be used to determine approximately the regulation of the transformer for any current and angle of phase difference.

The primary voltage is taken as 100 volts. Circles for secondary current 10 amperes, 5 amperes, and 0 amperes are drawn, and the values of  $E_2$  for 10 amperes, and the following angles of phase difference are shown:  $0^\circ$  lag,  $14.2^\circ$  lag, (the same as in VI., giving practically the same result. It is not quite the same because  $C_\mu$  is here neglected),  $37^\circ$  lag (power factor .80) and  $37^\circ$  lead.

The last shows how the secondary voltage may rise when working a leaky transformer on a circuit containing capacity.

I have just lately been able to do a test which completely confirms my theories.

The two coils before you form a leaky transformer which is very suitable for this purpose, and I drew out the diagram for them working in short circuit with the fixed coil as the primary some time ago.



Recently I was able to test them under the same conditions, and found the result to agree admirably as seen from the following table.

DRAWING.		TEST.
Frequency 50		51.3
$E_1$	70.4	68
$C_1$	1.15	1.15
$C_2$	.56	.56 (worked out from $W_1 - C_1^2 R_1$ )
$C_1^2 R_1 + C_2^2 R_2$	36.1	(36.1 not measured)
Watts, $W_1$	36.2	(36.1 lag wattmeter)
$\phi_1$	63.5°	62.8°

Note.—The frequency was 2.6% too high in the test and the voltage was 3.3% too low.

#### PARTICULARS OF EXPERIMENTAL TRANSFORMER.

COIL USED AS PRIMARY.		COIL USED AS SECONDARY.
Resistance ..	19.0 ohms.	35.0 ohms.
Wire ..	733 turns. No. 20.	820 turns. No. 22.
Co-efficient of Self-Induction	.250 Henry.	.334 Henry.
Co-efficient of Mutual Induction		
		.171. Coils close together.
Leakage co-efficient = $\frac{\sqrt{L_1 L_2}}{M} = 1.69.$		

$M$  is obtained, the mean of three experimental values for  $p M C\mu$ .

The leakage co-efficient obtained experimentally as the ratio of  $-p L_1 C\mu$  to  $p M C_2$  came out practically the same as the above, viz., 1.67.

$$L_1^1 = (.250 - .171 \times \frac{733}{820}) = (.250 - .153) = .097 \text{ Henry.}$$

$$L_2^1 = (.334 - .171 \times \frac{820}{733}) = (.334 - .191) = .143 \text{ Henry.}$$

On open circuit, with 68 volts on the primary, the primary current  $C_{0\mu}$  is .85 amperes, and  $\phi_1$  is 76°.

On short circuit with 70.4 volts on the primary, the primary current  $C_1 = 1.15$ , and  $C\mu$  is reduced to .59 amperes,  $C_2$  being .56 amperes.

#### CONCLUSIONS.

I think that I have proved conclusively that:—

- (1) In a leaky transformer  $E_{t2} = -p L_1 C\mu \times k$  and not  $-p L_1 C\mu \times \frac{n_2}{n_1}$  (as given in many books, including Kapp.)
- (2) In a leaky transformer the primary load current =  $C_2 \frac{n_2}{n_1}$  and not  $C_2 k$ .

- (3) In a leaky transformer  $C\mu$  should be reduced on load, as  $p L_1 C\mu$  is reduced.
- (4) It is quite unnecessary to introduce any lines in the diagrams to show magnetic fluxes, which lines only complicate the diagram.

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LIST OF BOOKS CONSULTED.

Transformers—Kapp.  
The Alternate Current Transformer—Fleming.  
Alternate Current Transformer—Baum.  
Alternating Currents—Franklin and Williamson.  
Alternating Currents and Alternating Current Machinery—Jackson  
Alternating Currents—Lamb.  
Practical Alternating Current—Smith.  
Alternating Currents—Hay.

